

NASA TECHNICAL
MEMORANDUM

NASA TM X- 64600

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AN ALGORITHM FOR SOLVING THE POPOV
CRITERION APPLIED TO SAMPLED DATA SYSTEMS

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April 2, 1971

NASA

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1. REPORT NO. NASA TM X-64600		2. GOVERNMENT ACCESSION NO.		3. RECIPIENT'S CATALOG NO.	
4. TITLE AND SUBTITLE An Algorithm for Solving the Popov Criterion Applied to Sampled Data Systems				5. REPORT DATE April 2, 1971	
				6. PERFORMING ORGANIZATION CODE	
7. AUTHOR(S)				8. PERFORMING ORGANIZATION REPORT #	
9. PERFORMING ORGANIZATION NAME AND ADDRESS George C. Marshall Space Flight Center Marshall Space Flight Center, Alabama 35812				10. WORK UNIT NO.	
				11. CONTRACT OR GRANT NO.	
12. SPONSORING AGENCY NAME AND ADDRESS National Aeronautics and Space Administration Washington, D. C. 20546				13. TYPE OF REPORT & PERIOD COVERED Technical Memorandum	
				14. SPONSORING AGENCY CODE	
15. SUPPLEMENTARY NOTES Prepared by Astrionics Laboratory, Science and Engineering					
16. ABSTRACT A systematic means is provided for analytically solving the Popov stability criterion that was developed by Tsypkin for application to sampled data systems. It is developed in a manner that is amenable to solution by a digital computer, regardless of the order of the polynomials defining the numerator and denominator of the z-transform of the transfer function representing the linear part of the plant.					
17. KEY WORDS Stability analysis Nonlinear stability theory Popov criterion Sampled data theory				18. DISTRIBUTION STATEMENT STAR Announcement 	
19. SECURITY CLASSIF. (of this report) Unclassified		20. SECURITY CLASSIF. (of this page) Unclassified		22. PRICE \$3.00	
				21. NO. OF PAGES 35	

TABLE OF CONTENTS

	Page
SUMMARY	1
INTRODUCTION	1
DESCRIPTION OF PROPOSED ALGORITHM	2
DERIVATION	4
EXAMPLE	6
CONCLUSION	7
APPENDIX A: DERIVATION OF RECURSION RELATIONSHIP.	9
APPENDIX B: DIGITAL COMPUTER PROGRAM DESCRIPTION	11
REFERENCES.	30

LIST OF ILLUSTRATIONS

Figure	Title	Page
1.	Model of class of nonlinear sampled data systems	2
2.	Digital computer logic diagram	5
3.	Model of integral-control servo system	6

LIST OF TABLES

Table	Title	Page
1.	Transfer Function Coefficients for Integral-Control Servo System	7
2.	Stability Limit on K	7

AN ALGORITHM FOR SOLVING THE POPOV¹ CRITERION APPLIED TO SAMPLED DATA SYSTEMS

SUMMARY

A systematic means is provided for analytically solving the Popov stability criterion that was developed by Tsytkin [1] for application to sampled data systems. It is developed in a manner that is amenable to solution by a digital computer, regardless of the order of the polynomials defining the numerator and denominator of the z-transform of the transfer function representing the linear part of the plant.

INTRODUCTION

Lindorff [2] has presented a derivation of Popov's stability criterion applied to sampled data systems which follows the development prepared by Tsytkin [1]. It provides a criterion for establishing sufficient conditions for absolute stability of a class of nonlinear sampled data systems described in Figure 1. The system must be capable of being separated into a single-valued nonlinearity represented by $\phi(\cdot)$ and a linear plant represented by $G(s)$. If this system meets the conditions,

$$\phi(0) = 0 \quad , \quad (1)$$

$$K > \frac{\phi(\sigma)}{\sigma} > 0 \text{ for } \sigma \neq 0 \quad , \quad (2)$$

and $G(s)$ represents a stable plant with at most one free integrator, then the Tsytkin version of the Popov stability criterion is met if

$$\operatorname{Re} \left\{ G(z) \mid z = e^{i\omega T} \right\} > -\frac{1}{K} \quad , \quad (3)$$

1. A portion of this report was presented as Paper No. G5 at the Third Annual Southeastern Symposium on System Theory at the Georgia Institute of Technology, Atlanta, Georgia, on April 5-6, 1971.

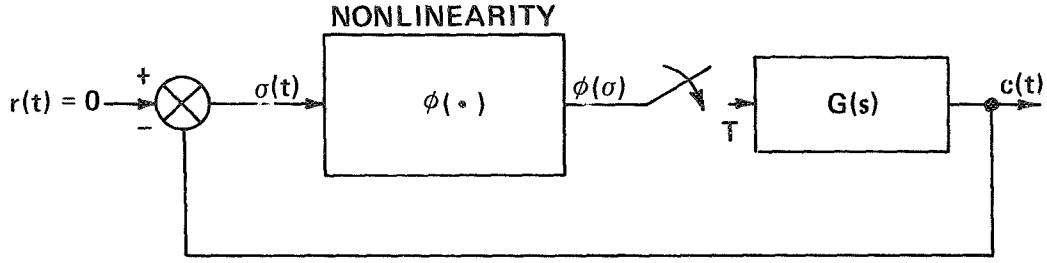


Figure 1. Model of class of nonlinear sampled data systems.

where $z \equiv e^{Ts}$, T is the sampling period, and $G(z)$ represents the z -transform of $G(s)$ which is evaluated at $z = e^{i\omega T}$, i.e., on the unit circle in the z -plane.

As with continuous systems, the stability requirement on the linear part of the plant can be weakened to include open-loop unstable plants [3]. This is shown through a coordinate transformation,

$$G_{tr}(s) = \frac{G(s)}{1 + \epsilon G(s)} \quad , \quad (4)$$

where gain ϵ is chosen so that the poles of the transformed system open-loop transfer $G_{tr}(s)$ lie within the unit circle in the z -plane [2]. This is equivalent to requiring the nonlinear function $\phi(\sigma)$ to lie within the sector $[\epsilon, k]$ for the original system to be absolutely stable.

In determining the upper bound K of the sector, the real part of $G(z)$ must be determined in some manner. This may be done by plotting $G(z)$ in a complex plane or by analytically determining the real part of $G(z)$ and then finding its minimum value. Either procedure becomes tedious as the order of a system under investigation increases.

DESCRIPTION OF PROPOSED ALGORITHM

A systematic means for analytically solving the criterion of equation (3) is provided, as a subsequent development to one suggested earlier by the author in a more primitive form [4]. It is developed in a manner that is particularly amenable to solution by a digital computer or electronic calculator, regardless of the order of the polynomials describing the numerator and

denominator of $G(z)$. It requires only that $G(z)$ be expressed in terms of two polynomials in z ; i.e.,

$$G(z) = \frac{\sum_{j=0}^n \gamma_j z^j}{\sum_{j=0}^d \eta_j z^j} . \quad (5)$$

The criterion of equation (3) may be recast in the form

$$\frac{N_1 D_1 + (1 - \beta^2) N_2 D_2}{D_1^2 + (1 - \beta^2) D_2^2} > -\frac{1}{K} , \quad (6)$$

where

$$\begin{aligned} N_1 &= \sum_{j=0}^n \gamma_j X_j , \\ N_2 &= \sum_{j=0}^n \gamma_j Y_j , \\ D_1 &= \sum_{j=0}^d \eta_j X_j , \\ D_2 &= \sum_{j=0}^d \eta_j Y_j , \end{aligned} \quad (7)$$

and

$$\beta = \cos \omega T . \quad (8)$$

The terms X_j and Y_j are developed by means of the recursive relationships,

$$\begin{aligned} X_{j+1} - 2\beta X_j + X_{j-1} &= 0 , \\ Y_{j+1} - 2\beta Y_j + Y_{j-1} &= 0 , \end{aligned} \quad (9a)$$

where

$$X_0 = Y_1 = 1 \quad ,$$

$$Y_0 = 0 \quad ,$$

and

$$X_1 = \beta \quad ,$$

or by the alternate recursive relationships,

$$X_{j+1} = \beta X_j - (1 - \beta^2) Y_j \quad ,$$

$$Y_{j+1} = X_j + \beta Y_j \quad . \quad (9b)$$

See Appendix A for the development of these relationships.

Figure 2 represents a logic diagram demonstrating the ease with which a general program can be established to solve equation (6) for any plant meeting the conditions specified in the introduction. The linear portion of the plant may be of any order as long as it can be expressed in terms of a z-domain open-loop transfer function. See Appendix B for computer program.

DERIVATION

The derivation of equation (6) from equation (3) is brief. Defining \bar{z} as

$$\bar{z} \equiv e^{i\omega T} \quad (10)$$

it follows that \bar{z}^j may be expressed as

$$\bar{z}^j = X_j + i Y_j \sqrt{1 - \beta^2} \quad . \quad (\text{See Appendix A.}) \quad (11)$$

It is apparent from the definition of equation (10) that

$$\begin{aligned} X_0 &= 1 \quad , & Y_0 &= 0 \quad , \\ X_1 &= \beta \quad , & Y_1 &= 1 \quad . \end{aligned} \quad (12)$$

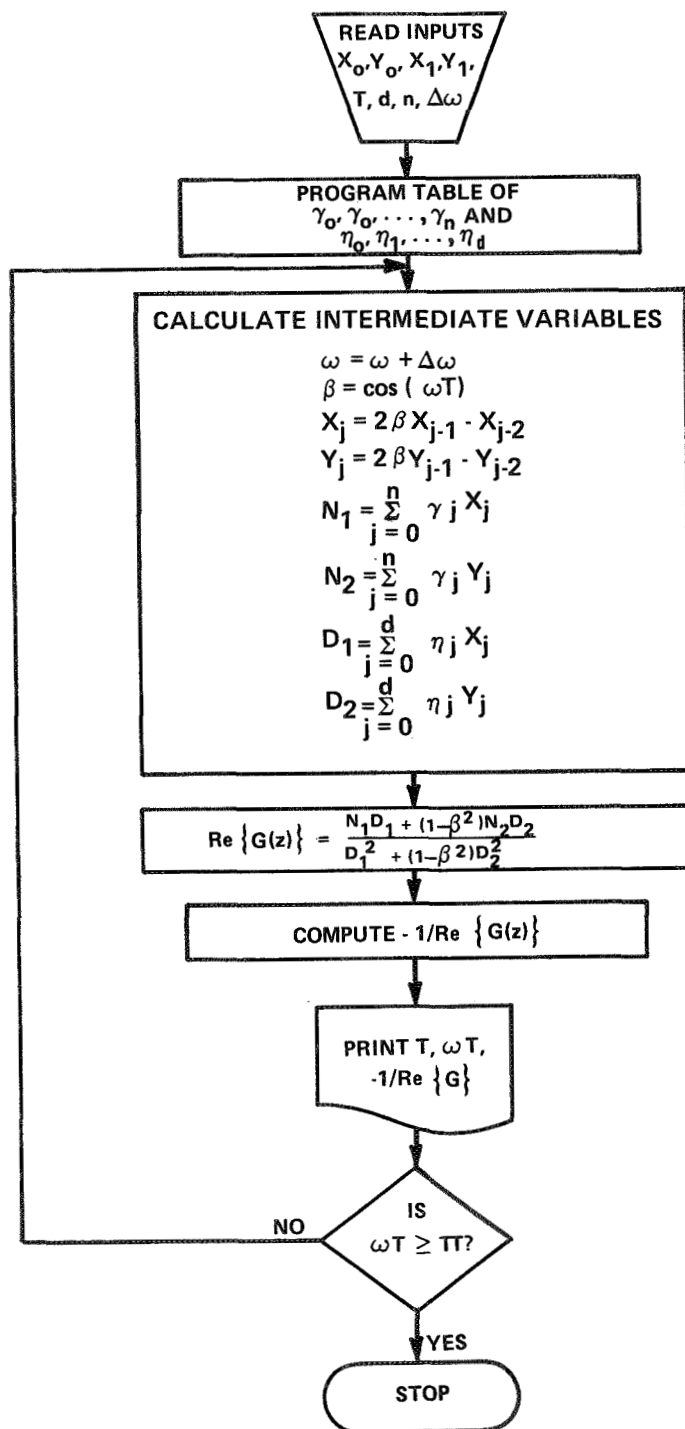


Figure 2. Digital computer logic diagram.

One may use equations (12) and the recursion relations of equations (9) to generate values of X_j and Y_j for $j > 1$. Then, equation (11) may be substituted into equation (5), letting $z = \bar{z}$,

$$G(z) \Big|_{z=e^{i\omega T}} = \frac{\sum_{j=0}^n \gamma_j X_j + i \sum_{j=0}^n \gamma_j Y_j \sqrt{1 - \beta^2}}{\sum_{j=0}^d \eta_j X_j + i \sum_{j=0}^d \eta_j Y_j \sqrt{1 - \beta^2}} . \quad (13)$$

Taking the real part of equation (13) yields the left half side of equation (6).

EXAMPLE

Several examples show how the proposed algorithm is implemented. Consider an integral control servo system [5] shown in Figure 3. The corresponding z -transform of the linear portion of the plant may be put in the form of equation (5) and is described in Table 1. By using Table 1 as an input to the digital program, the left side of inequality (6) may be solved as ωT varies in value from zero to π for selected values of T . The results are printed out, and the minimum value of

$$- 1/\text{Re} \left\{ G(z) \Big|_{z=e^{i\omega T}} \right\}$$

is selected as the maximum value which K may possess. The results, reproduced in Table 2, agree with those of Reference 5. The same algorithm has been used to find stability limits on K for two other examples of Reference 5 (Bertram's example and the integral-control example with digital compensation added). The results have been identical.

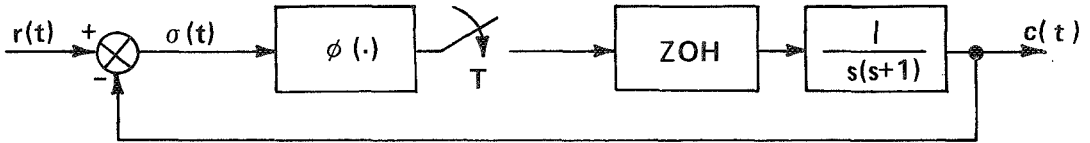


Figure 3. Model of integral-control servo system.

TABLE 1. TRANSFER FUNCTION COEFFICIENTS
FOR INTEGRAL-CONTROL SERVO SYSTEM

j	γ_j	η_j
0	$1 - e^{-T} - Te^{-T}$	e^{-T}
1	$T - 1 + e^{-T}$	$-(1 + e^{-T})$
2	0	1

TABLE 2. STABILITY LIMIT ON K

Sampling Period T					
	1	2	3	4	5
K<	0.666	0.5	0.4	0.333	0.286

CONCLUSION

A general computer program can be developed to determine the sufficient condition for stability of a class of nonlinear sampled data systems in terms of a sector within which the nonlinearity must lie. The criterion used in developing the computer algorithm is Tsypkin's development of the Popov stability criterion. Work is now in progress to investigate a similar algorithm for Jury's and Lee's [5] less restrictive stability criterion,

$$\operatorname{Re}\{G(z)[1 + q(z - 1)]\} + \frac{1}{K} - \frac{K'|q|}{2} \left| (z - 1)G(z) \right|^2 \leq 0, \quad (14)$$

where

$$-\infty \leq \frac{d\phi(\sigma)}{d\sigma} < K',$$

and q is a real number.

APPENDIX A²

DERIVATION OF RECURSION RELATIONSHIP

The two sets of recursive relations, equations (9a) and equations (9b), are developed from the definition of \bar{z} ,

$$\bar{z} \equiv e^{i\omega T} = \cos \omega T + i \sin \omega T \quad . \quad (10)$$

Letting

$$\beta = \cos \omega T \quad , \quad (8)$$

\bar{z} may be rewritten as

$$\bar{z} = \beta + i\sqrt{1 - \beta^2} \quad . \quad (A-1)$$

If \bar{z} is raised to the j^{th} power, it may be written as

$$\bar{z}^j = e^{ij\omega T} = \cos j\omega T + i \sin j\omega T \quad . \quad (A-2)$$

Two identities [6] for $\cos j\omega t$ and $\sin j\omega t$ may be rewritten in terms of β :

$$\cos j\omega T = 2^{j-1} \beta^j - \sum_{p=0}^{\frac{1}{2}(j-3)} (-1)^p \binom{j-p-2}{p} \left(\frac{j}{p+1} \right) 2^{j-2p-3} \beta^{j-2p-2} \quad . \quad (A-3)$$

and

$$\sin j\omega T = \sqrt{1 - \beta^2} \sum_{p=0}^{\frac{1}{2}(j-1)} (-1)^p \binom{j-p-1}{p} 2^{j-2p-1} \beta^{j-2p-1} \quad , \quad (A-4)$$

where p is an integer. If $\sin j\omega T$ and $\cos j\omega T$ are defined as

2. I am indebted to Mr. Hans Hosenthien, Chief, R&D Analysis Office, Astrionics Laboratory, NASA, Marshall Space Flight Center, for the development of the alternate set of recursive relations of equations (9b).

$$\cos j\omega T \equiv X_j$$

and

$$\sin j\omega T \equiv Y_j \sqrt{1 - \beta^2} \quad , \quad (A-5)$$

examination of equations (A-3) and (A-4) shows that X_j and Y_j are polynomials in β . If equations (A-5) are substituted into equation (A-2), the resulting expression becomes of the form

$$\bar{z}^j = X_j + iY_j \sqrt{1 - \beta^2} \quad . \quad (11)$$

The recursive relations of equations (9a) or (9b) may be derived by writing \bar{z}^{j+1} as

$$\begin{aligned} \bar{z}^{j+1} &= \bar{z}^j \bar{z} \\ &= \left(X_j + iY_j \sqrt{1 - \beta^2} \right) \left(\beta + i\sqrt{1 - \beta^2} \right) \\ &= \left[X_j \beta - Y_j (1 - \beta^2) \right] + i \left[\left(X_j + Y_j \beta \right) \sqrt{1 - \beta^2} \right] \\ &= X_{j+1} + iY_{j+1} \quad , \end{aligned}$$

yielding the recursive relations

$$X_{j+1} = X_j \beta - Y_j (1 - \beta^2)$$

and

$$Y_{j+1} = X_j + Y_j \beta \quad . \quad (9b)$$

From the definition of \bar{z} ,

$$X_0 = Y_1 = 1 \quad , \quad Y_0 = 0 \quad , \quad X_1 = \beta \quad . \quad (A-7)$$

Using equations (9b), one may readily derive equations (9a).

APPENDIX B

DIGITAL COMPUTER PROGRAM DESCRIPTION

The Popov-Tsytkin digital program was written in Fortran IV, and computer runs were made on an SEL 840 digital computer. Using the Tsytkin version of Popov's criterion, the program calculates and prints the negative reciprocal of the real part of the z-domain transfer function representing the linear portion of the plant. Values of the negative reciprocal are calculated for incremental values of the independent argument ωT and scanned; the smallest value is selected as the upper bound of the nonlinear (or Popov) sector. The following mathematical model is described by equations (3) and (6) through (9a) and is the basis for all program calculations (see program symbol table for relations of mathematical symbols and program symbols).

Program Symbol Table

<u>Math Symbol</u>	<u>Program Symbol</u>	<u>Description</u>
X_0	X(1)	Initial value for X.
Y_0	Y(1)	Initial value for Y.
X_1	X(2)	Initial condition for X.
Y_1	Y(2)	Initial condition for Y.
	AIN	Increment denominator which is divided into RA to form the omega increment value.
	RA	Range value of omega.
	WI	Initial omega.
T	T	T parameter which should be set to 1.0 and associated with ωT limit values.
$\Delta\omega$	DW	Delta omega which value is used if omega equals 0.

<u>Math Symbol</u>	<u>Program Symbol</u>	<u>Description</u>
d	ID	Upper limit on D_1 and D_2 summation and has value $d + 1$.
n	N	Upper limit on N_1 and N_2 summation and has value $n + 1$.
K'	KP	Program flag which when set to 0 causes a Popov-Tsyarkin run to be made. This relates to a program modification now under development to solve equation (14).

See program inputs for all other values.

Program Inputs

Card 1 — Reads $X(1)$, $Y(1)$, and $Y(2)$ under a 4F15.8 format. These are the initial X and Y values.

Card 2 — Reads AIN , RA , and WI which are the increment value of ω , the range of ω , and the initial ω . Reads under a 4F15.8 format.

Card 3 — Reads T under an F15.8 format.

Card 4 — Reads DW , the delta ω value, under an F15.8 format.

Card 5 — Reads ID , the value for d , under an I5 format.

Card 6 — Reads N , the value for n , under an I5 format.

Card 7 — Reads KP , the value which when set to 0 causes the program to make a Popov-Tsyarkin run using the initial equations to compute $\operatorname{Re}\{G(Z)\}$ and its negative reciprocal. When $KP = 1$, the program omits this phase and computes a K' equation set; where $KP = 2$, the above options are omitted and the program solves a quadratic K equation. For the Popov-Tsyarkin option always set $KP = 0$.

Cards 8 through 11 contain the inputs involving options other than the Popov-Tsyarkin equation solutions that were part of the program study and are included in the present version of the program. These inputs are included here only for the programmers information.

Card 8 — Reads IQ, the number of Q elements, under an I5 format.

Card 9 — Reads IK, the number of K elements, under an I5 format.

Card 10 — Reads array Q, containing all values of q, under a 4F15.8 format.

Card 11 — Reads array AK, containing all K elements, under a 4F15.8 format.

Sample Run Case

The following sample case was run successfully on an SEL 840 digital computer. The appropriate gamma and eta functions listed in Table 1 of the computer printout were used in the Popov-Tsyarkin program with the input data listed here and described under the card input data. The input data will always be followed by the printed variable values for T , ωT , $\text{Re} \{G(Z)\}$, and $K < -1/\text{Re} \{G(Z)\}$.

```

C* * * * * P0P0V - ISYPKIN PROGRAM * * * * * 1
  DIMENSION Q(20),AK(20) 2
  DIMENSION X(20),Y(20),G(20),E(20) 3
  DIMENSION IALA(20) 4
  TRACE 9999 5
  READ (4,10) X(1),Y(1),Y(2) 6
  READ (4,10) AIN,RA,WI 7
  READ (4,11) T 8
  READ (4,11) DW 9
  READ (4,12) ID 10
  READ (4,12) N 11
  READ (4,12) KP 12
  READ (4,12) IQ 13
  READ (4,12) IK 14
  READ (4,10) (Q(I),I=1,IQ) 15
  READ (4,10) (AK(I),I=1,IK) 16
  READ (4,120) IALA 17
10  FORMAT (4F15.8) 18
11  FORMAT (F15.8) 19
12  FORMAT (I5) 20
120  FORMAT (20A4) 21
  WRITE (5,120) IALA 22
  WRITE (5,20) 23
20  FORMAT(5X,10HINPUT DATA//) 24
  WRITE (5,21) X(1),Y(1),Y(2) 25
21  FORMAT (10X,7HX(1) = ,F10.5,3X,7HY(1) = ,F10.5,3X,7HY(2) = ,F10.5) 26
  WRITE (5,26) AIN,RA,WI 27
26  FORMAT (10X,18H0MEGA INCREMENT = ,F10.5,3X,14H0MEGA RANGE = ,F10.5 28
1,3X,16H0MEGA INITIAL = ,F10.5) 29
  WRITE (5,22) ID 30
22  FORMAT (10X,4HD = ,I5) 31
  WRITE (5,23) N 32
23  FORMAT (10X,4HN = ,I5) 33
  WRITE (5,24) T 34
24  FORMAT (10X,4HT = ,F15.8) 35
  WRITE (5,27) DW 36
27  FORMAT (10X,14HDELTA 0MEGA = ,F15.8) 37
  WRITE (5,500) KP,IQ,IK 38
500  FORMAT (10X,5HKP = ,I5,5X,5HIQ = ,I5,5X,5HIK = ,I5) 39
  WRITE (5,501) (I,Q(I),I=1,IQ) 40
501  FORMAT (10X,2HQ(,I2,4H) = ,F10.5) 41
  WRITE (5,502) (I,AK(I),I=1,IK) 42
502  FORMAT (10X,2HK(,I2,4H) = ,F10.5) 43
  W=0.0 44
C**FORM GAMMA AND ETA FUNCTION 45
  E1=EXP(-T) 46
  G(1)=(1.0-E1)-T*E1 47
  G(2)=T*(1.0-E1) 48
  E(1)=E1 49
  E(2)=-(1.0+E1) 50
  E(3)=1.0 51
C**FORM REAL ROOT FUNCTION 52
  IN=AIN 53
  IC=IN+1 54
  VA=RA/AIN 55
  IF (KP=1) 4010,401,401 56
4010  CONTINUE 57

```

	WRITE (5,41)	58
41	FORMAT (11X,1HT,18X,2HWT,14X,8HRE(G(Z)),4X,23HK LESS THAN -1/RE(G(59
	1Z)))	60
401	CONTINUE	61
	IF (KP=1) 39,301,390	62
39	CONTINUE	63
	DØ 40 I=1,IC	64
	AI=I	65
	W=AI*VA*AI	66
	IF (W.EQ.0.0) W=DW	67
	WT=W*T	68
	B=CØS(WT)	69
	X(2)=B	70
	IF (ID.LT.3) GØ TØ 410	71
	DØ 30 J=3,ID	72
	X(J)=2.0*B*X(J-1)-X(J-2)	73
	Y(J)=2.0*B*Y(J-1)-Y(J-2)	74
30	CONTINUE	75
	CALL SUM (N,ID,IP,X,Y,E,G,AN1,AN2,D1,D2)	76
	GØ TØ 300	77
410	CONTINUE	78
	CALL SUM (N,ID,IP,X,Y,E,G,AN1,AN2,D1,D2)	79
300	CONTINUE	80
	ANUM=AN1*D1+(1.0-B**2)*AN2*D2	81
	DEN=D1*D1+(1.0-B**2)*D2*D2	82
	RE=ANUM/DEN	83
	REC=-1.0/RE	84
	WRITE (5,31) T,WT,RE,REC	85
31	FORMAT (3X,E15.8,5X,E15.8,5X,E15.8,5X,E15.8)	86
40	CONTINUE	87
	GØ TØ 4200	88
301	CONTINUE	89
	DØ 380 IJ=1,IK	90
	DØ 380 II=1,IQ	91
	DØ 380 I=1,IC	92
	AI=I	93
	W=WI+VA*AI	94
	IF (W.EQ.0.0) W=DW	95
	WT=W*T	96
	B=CØS(WT)	97
	X(2)=B	98
	BM1=1.0-B	99
	BSQ=1.0-B*B	100
	IF (ID.LT.3) GØ TØ 4100	101
	DØ 3000 J=3,ID	102
	X(J)=2.0*B*X(J-1)-X(J-2)	103
	Y(J)=2.0*B*Y(J-1)-Y(J-2)	104
3000	CONTINUE	105
	CALL SUM (N,ID,IP,X,Y,E,G,AN1,AN2,D1,D2)	106
	GØ TØ 3111	107
4100	CONTINUE	108
	CALL SUM (N,ID,IP,X,Y,E,G,AN1,AN2,D1,D2)	109
3111	CONTINUE	110
	RK=1.0/AK(IJ)	111
	DBV=AN1*D1+BSQ*AN2*D2	112
	QB=1.0-Q(II)*BM1	113
	BQ=G(II)*BSQ	114
	DN=AN1*D2-AN2*D1	115

	DB=D1*D1+BSQ*D2*D2	116
	TOP=DBN*QB*BQ*DN*RK*DB	117
	AQB=Q(II)*BM1	118
	BN=AN1*AN1+BSQ*AN2*AN2	119
	BOT=AQB*BN	120
	AKP=TOP/BOT	121
	WRITE (5,385) IJ,AK(IJ),II,Q(II),WT,AKP	122
383	FORMAT (3X,2HK(,I2,4H) = ,E15.8,2X,2HQ(,I2,4H) = ,E15.8,2X,5HWT =	123
	1,E15.8,2X,7HK PR = ,E15.8)	124
380	CONTINUE	125
	G0 T0 4200	126
390	CONTINUE	127
	D0 399 II=1,IQ	128
	D0 399 I=1,IC	129
	AI=I	130
	W=WI+VA*AI	131
	IF (W,EQ,0,0) W=DW	132
	WT=W*T	133
	B=COS(WT)	134
	X(2)=B	135
	BM1=1.0-B	136
	BSQ=1.0-B*B	137
	IF (ID,LT,3) G0 T0 4101	138
	D0 3001 J=3,ID	139
	X(J)=2.0*B*X(J-1)-X(J-2)	140
	Y(J)=2.0*B*Y(J-1)-Y(J-2)	141
3001	CONTINUE	142
	CALL SUM (N,ID,IP,X,Y,E,G,AN1,AN2,D1,D2)	143
	G0 T0 3112	144
4101	CONTINUE	145
	CALL SUM (N,ID,IP,X,Y,E,G,AN1,AN2,D1,D2)	146
3112	CONTINUE	147
	AA=Q(II)*BM1*(AN1*AN1+BSQ*AN2*AN2)	148
	BB=- (AN1*D1+BSQ*AN2*D2)*(1.0-Q(II)*BM1)	149
	CC=- (D1*D1+BSQ*D2*D2)	150
	DISC=SQRT(BB*BB-4.0*AA*CC)	151
	AK1=(-BB+DISC)/(2.0*AA)	152
	AK2=(-BB-DISC)/(2.0*AA)	153
	WRITE (5,3990) (II,Q(II),AK1,AK2,WT)	154
3990	FORMAT (3X,2HQ(,I2,4H) = ,E15.8,5X,5HK1 = ,E15.8,5X,5HK2 = ,E15.8,	155
	15X,5HWT = ,E15.8)	156
399	CONTINUE	157
4200	CONTINUE	158
	WRITE (5,42)	159
42	FORMAT (25X,14HCASE COMPLETED)	160
	WRITE (5,43)	161
43	FORMAT (1H1)	162
	G0 T0 1	163
	END	164
PATH		
00000	Q	
00050	AK	
00120	X	
00170	Y	
00240	G	
00310	E	
00360	IALA	
UNRF	\$9999	

00405	\$1	03323	T\$1003
00576	\$10	SUBR	SUM
03251	AIN	03324	IP
03253	RA	03325	AN1
03255	WI	03327	AN2
00601	\$11	03331	D1
03257	T	03333	D2
03261	DW	01556	\$300
00604	\$12	03335	ANUM
03263	ID	03337	DEN
03264	N	03341	RE
03265	KP	03343	REC
03266	IQ	01661	\$31
03267	IK	03217	\$4200
03270	I	02457	\$380
03271	T\$1000	03347	IJ
00606	\$120	03350	II
00624	\$20	03351	BM1
00647	\$21	03353	BSQ
00704	\$26	02137	\$4100
00743	\$22	02112	\$3000
00757	\$23	02156	\$3111
00773	\$24	03355	RK
01010	\$27	03357	DBN
01036	\$500	03361	QB
01100	\$501	03363	BQ
01136	\$502	03365	DN
03272	W	03367	DB
03276	E1	03371	T0P
SUBR	EXP	03373	T\$2001
03300	T\$2000	03375	AQB
03304	IN	03377	BN
03305	IC	03401	B0T
03306	VA	03403	AKP
01266	\$4010	02430	\$385
01316	\$401	03202	\$399
01275	\$41	02732	\$4101
01330	\$39	02705	\$3001
01705	\$301	02751	\$3112
02502	\$390	03405	AA
01674	\$40	03407	BB
03310	AI	03411	T\$2002
03312	WT	03413	CC
03314	B	03415	DISC
SUBR	COS	SUBR	SQRT
01537	\$410	03421	AK1
01512	\$30	03423	AK2
03316	J	03156	\$3990
03321	T\$1001	03226	\$42
03322	T\$1002	03241	\$43

	SUBROUTINE SUM (N, ID, IP, X, Y, E, G, AN1, AN2, D1, D2)	1
	DIMENSION X(20), Y(20), G(20), E(20)	2
	TRACE 9999	3
	AN1=0.0	4
	AN2=0.0	5
	D1=0.0	6
	D2=0.0	7
	DØ 10 J=1, N	8
	AN1=G(J)*X(J)+AN1	9
	AN2=G(J)*Y(J)+AN2	10
10	CONTINUE	11
	DØ 11 J=1, ID	12
	D1=E(J)*X(J)+D1	13
	D2=E(J)*Y(J)+D2	14
11	CONTINUE	15
	RETURN	16
	END	17

00000	N	(DUMMY)
00001	ID	(DUMMY)
00003	X	(DUMMY)
00004	Y	(DUMMY)
00005	E	(DUMMY)
00006	G	(DUMMY)
00007	AN1	(DUMMY)
00010	AN2	(DUMMY)
00011	D1	(DUMMY)
00012	D2	(DUMMY)
UNRF	\$9999	
00115	\$10	
00212	J	
00213	I\$1000	
00214	I\$1001	
00215	I\$1002	
00176	\$11	

TABLE 1 04/19//1
INPUT DATA

```

X(1) = 1.00000 Y(1) = .00000 Y(2) = 1.00000
OMEGA INCREMENT = 314.00000 OMEGA RANGE = 3.14000 OMEGA INITIAL = .00000
D = 3
N = 2
T = 1.00000000
DELTA OMEGA = .00100000
KP = 0 IQ = 9 IK = 4
Q( 1) = .00010
Q( 2) = .20000
Q( 3) = .40000
Q( 4) = .60000
Q( 5) = .80000
Q( 6) = 1.00000
Q( 7) = 2.00000
Q( 8) = 10.00000
Q( 9) = 100.00000
K( 1) = .48990
K( 2) = .49000
K( 3) = 1.99000
K( 4) = 2.00000

      T      WT      RE(G(Z))      K LESS THAN -1/RE(G(Z))
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.10000000E 01      .30000000E+01      -.14984609E 01      .66735141E 00
.10000000E 01      .40000000E+01      -.14972657E 01      .66788411E 00
.10000000E 01      .50000000E+01      -.14957316E 01      .66856916E 00
.10000000E 01      .60000000E+01      -.14938602E 01      .66940668E 00
.10000000E 01      .70000000E+01      -.14916539E 01      .67039679E 00
.10000000E 01      .80000000E+01      -.14891153E 01      .67153967E 00
.10000000E 01      .90000000E+01      -.14862474E 01      .67283550E 00
.10000000E 01      .10000000E 00      -.14830535E 01      .67428449E 00
.10000000E 01      .11000000E 00      -.14795375E 01      .67588689E 00
.10000000E 01      .12000000E 00      -.14757034E 01      .67764296E 00
.10000000E 01      .13000000E 00      -.14715556E 01      .67955299E 00
.10000000E 01      .14000000E 00      -.14670989E 01      .68161730E 00
.10000000E 01      .15000000E 00      -.14623385E 01      .68383621E 00
.10000000E 01      .16000000E 00      -.14572796E 01      .68621012E 00
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.10000000E 01      .22000000E 00      -.14209955E 01      .70373201E 00
.10000000E 01      .23000000E 00      -.14140206E 01      .70720326E 00
.10000000E 01      .24000000E 00      -.14067994E 01      .71083339E 00
.10000000E 01      .25000000E 00      -.13993393E 01      .71462298E 00
.10000000E 01      .26000000E 00      -.13916477E 01      .71857267E 00
.10000000E 01      .27000000E 00      -.13837323E 01      .72268314E 00
.10000000E 01      .28000000E 00      -.13756009E 01      .72695505E 00
.10000000E 01      .29000000E 00      -.13672612E 01      .73138914E 00
.10000000E 01      .30000000E 00      -.13587212E 01      .73598614E 00
.10000000E 01      .31000000E 00      -.13499889E 01      .74074683E 00
.10000000E 01      .32000000E 00      -.13410722E 01      .74567201E 00
.10000000E 01      .33000000E 00      -.13319791E 01      .75076251E 00

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.10000000E 01	.36000000E 00	-.13037220E 01	.76703470E 00
.10000000E 01	.37000000E 00	-.12940036E 01	.77279540E 00
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.10000000E 01	.39000000E 00	-.12741652E 01	.78482757E 00
.10000000E 01	.40000000E 00	-.12640609E 01	.79110111E 00
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.10000000E 01	.45000000E 00	-.12119931E 01	.82508718E 00
.10000000E 01	.46000000E 00	-.12013220E 01	.83241627E 00
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.10000000E 01	.48000000E 00	-.11797791E 01	.84761631E 00
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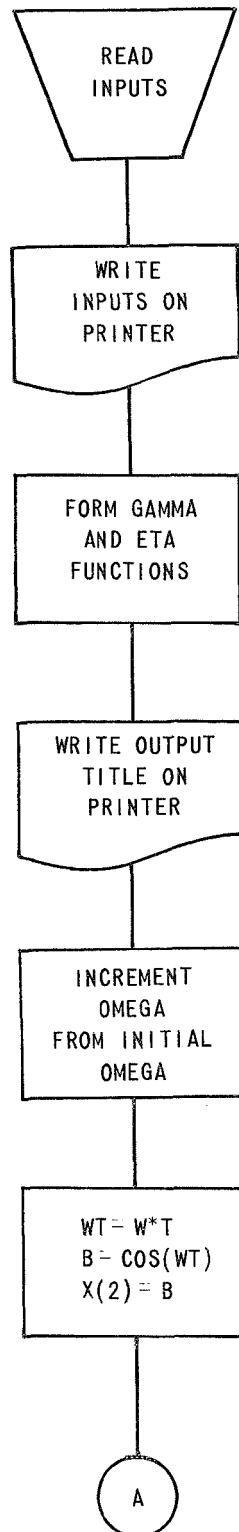
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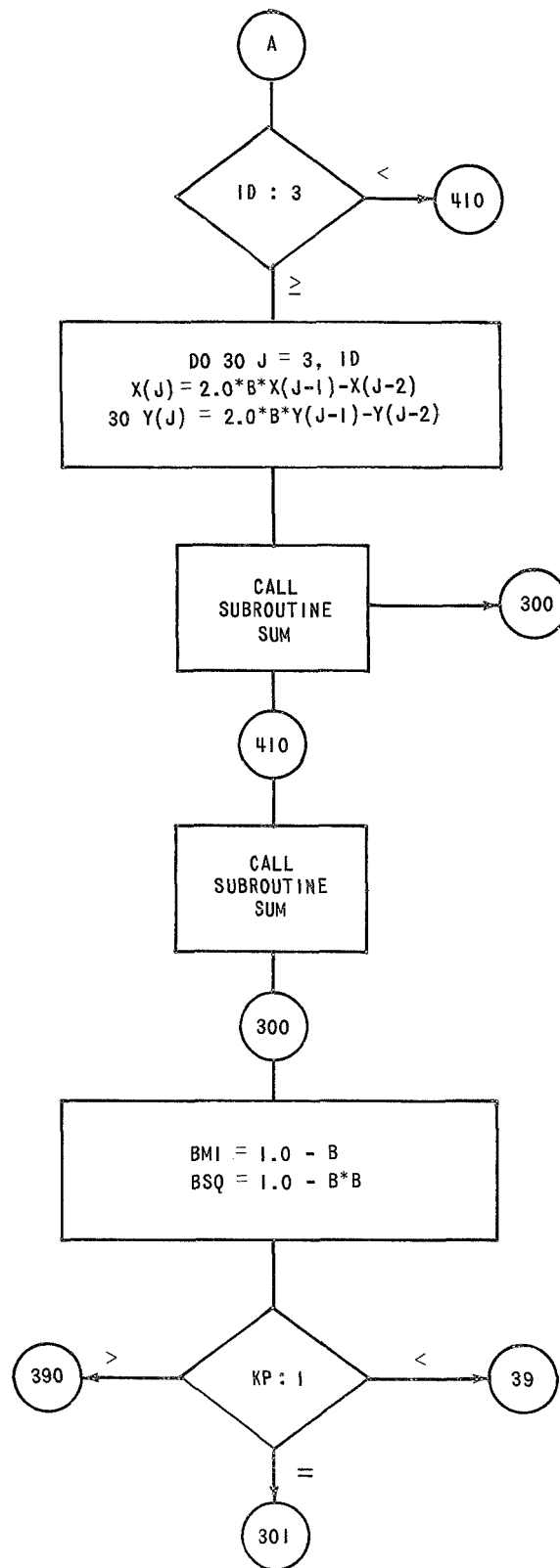
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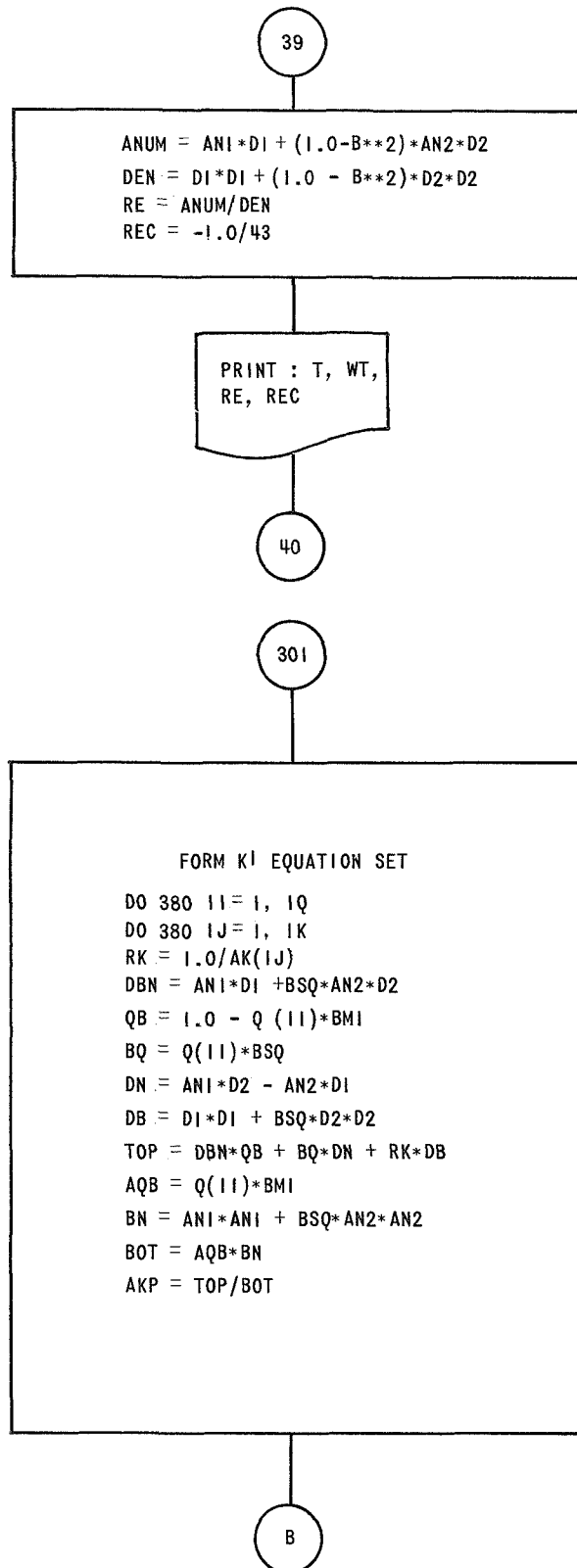
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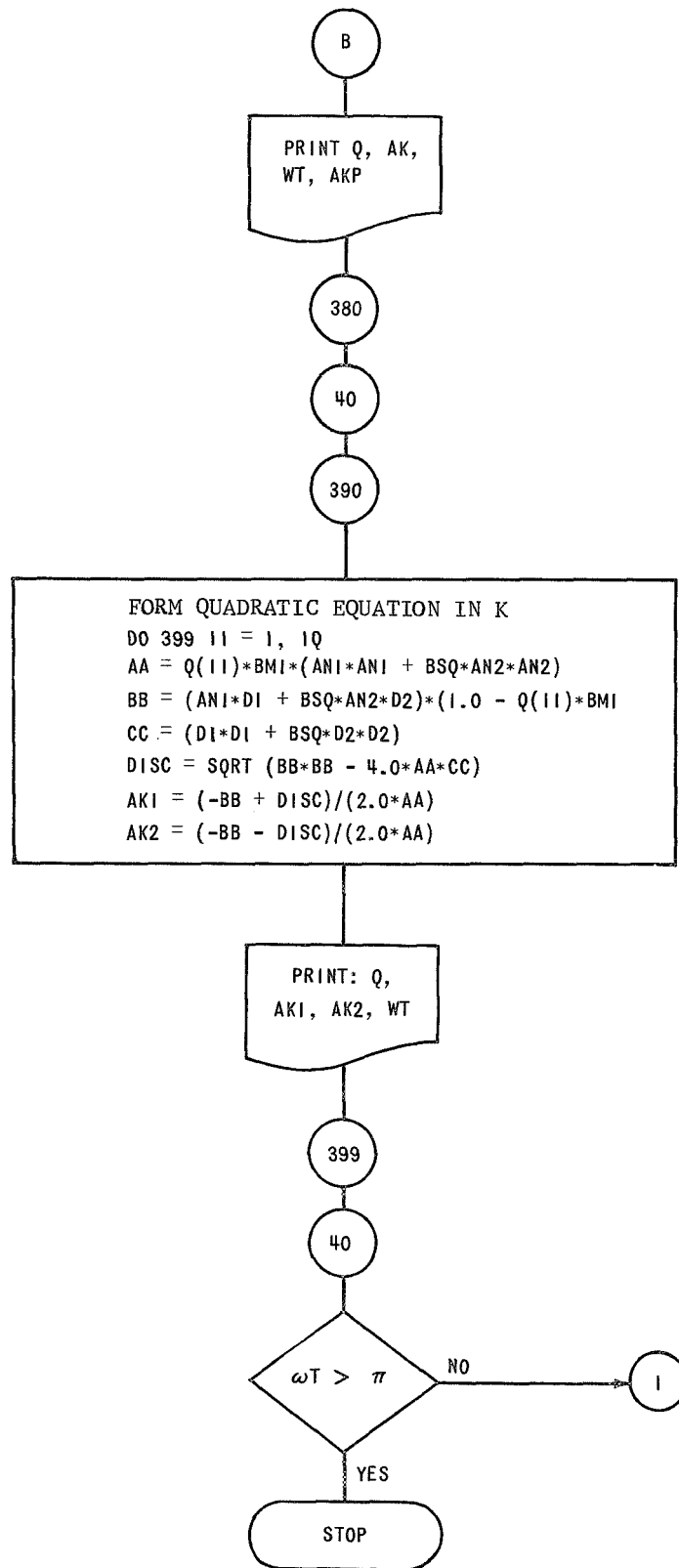
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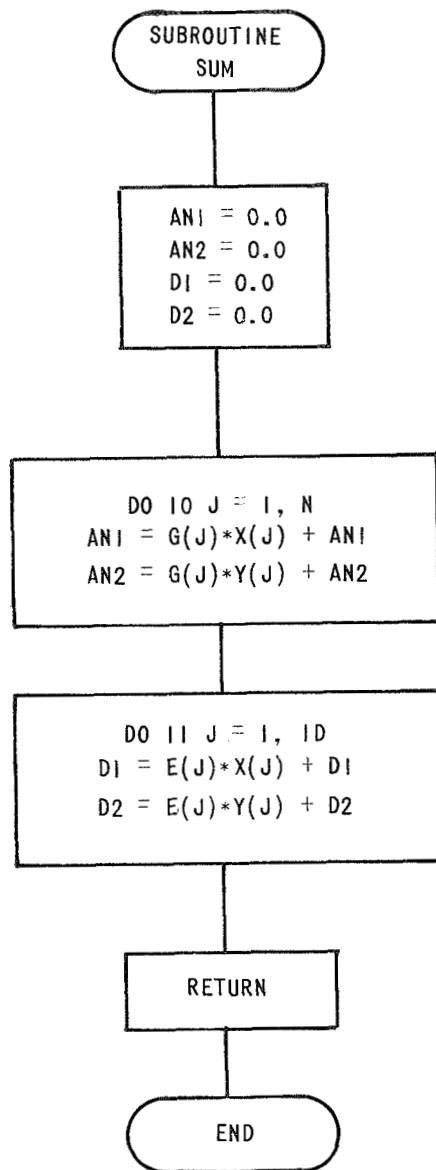
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FLOW DIAGRAM











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AN ALGORITHM FOR SOLVING THE POPOV CRITERION APPLIED TO SAMPLED DATA SYSTEMS

By S. M. Seltzer

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